

# Soft-Information-Based Joint Network-Channel Coding for the Two-Way Relay Channel

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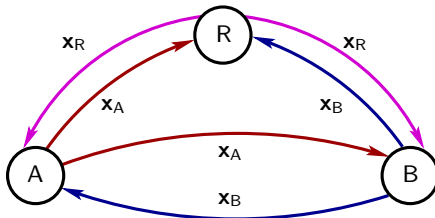


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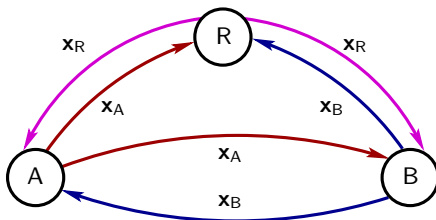
- 1 Two-way relay channel
- 2 Proposed scheme
- 3 Simulation results
- 4 Conclusions and outlook

- **Two-way relay channel (TWRC)**
  - ▶ A and B exchange independent messages
  - ▶ relay R supports transmission  $A \leftrightarrow B$
  - ▶ **large variety of settings and coding strategies**



- **Goal:** devise a coding scheme for the TWRC with
  - ▶ excellent **performance**
  - ▶ low **complexity** and **delay**
  - ▶ considering practical constraints

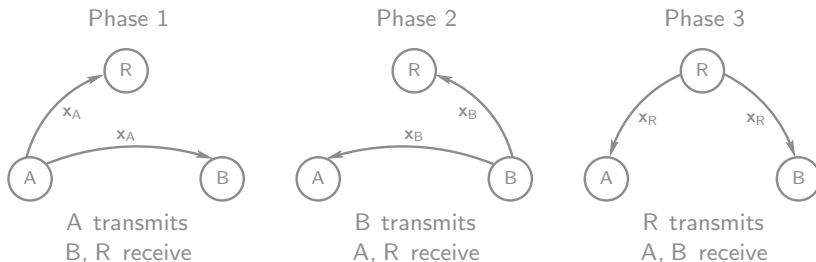
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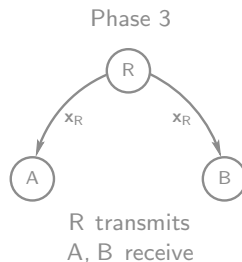
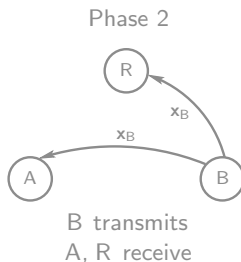
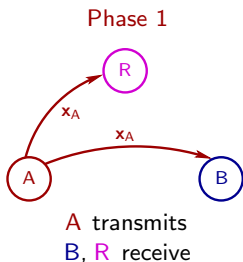
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  - ▶ easy to implement in practice
  - ▶ network coding increases spectral efficiency



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  - ▶ performs well even if relay can not decode

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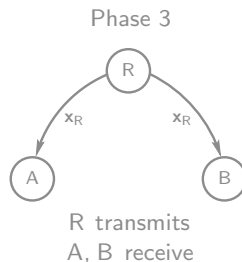
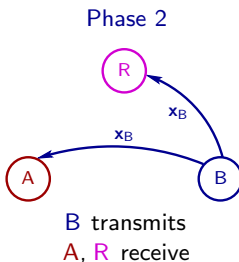
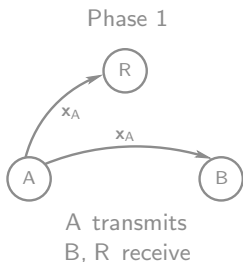
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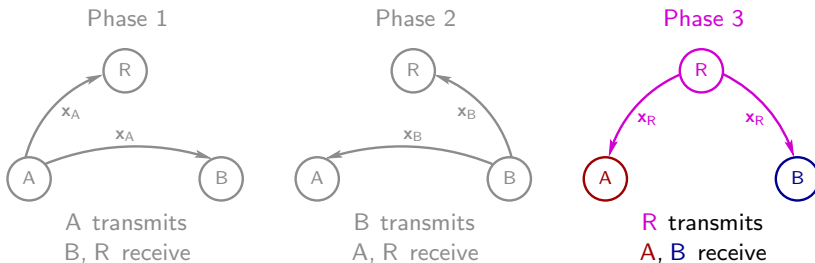
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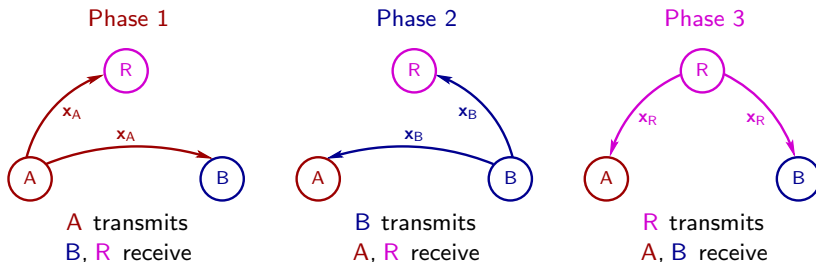
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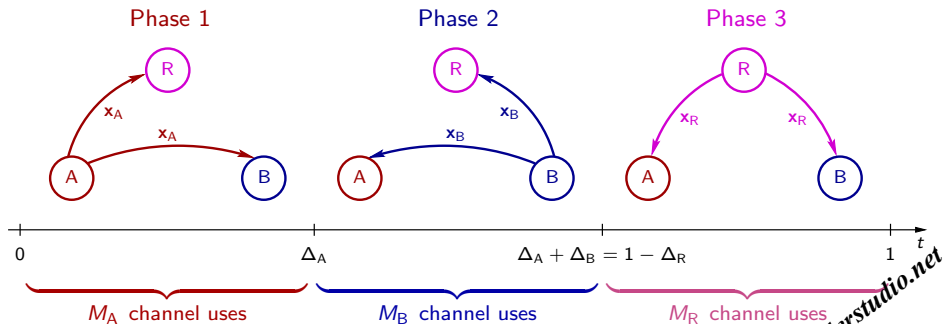
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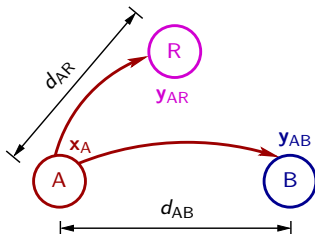
# Notation

- $M_i$  channel uses per node,  $\sum_i M_i = M$ ,  $i \in \{A, B, R\}$
- Transmit time  $\Delta_i = M_i/M$  per node
- User A (B) transmits  $K_A$  ( $K_B$ ) bits of information
- Rates  $R_A = K_A/M_A$ ,  $R_B = K_B/M_B$



# Channel Model

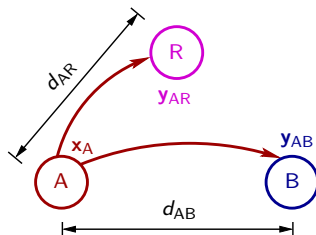
- Assume **Gaussian channels**:  $\mathbf{y}_{ij} = d_{ij}^{-n/2} \mathbf{x}_i + \mathbf{w}_{ij}$



- Assumptions
  - power constraint:  $E\{\|\mathbf{x}_i\|_2^2\} / M_i = 1$
  - noise statistics:  $\mathbf{w}_{ij} \sim \mathcal{CN}(\mathbf{0}, N_0 \mathbf{I})$
  - SNR  $\gamma_{ij} = d_{ij}^{-n} / N_0$ , with  $n = 3.52$

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- **Information-theoretic limits** [KimMitTar, TrIT08]
  - ▶ outer bound of capacity region
  - ▶ achievable rates for DF-based transmission
- Compute sum-rate **optimal transmit time allocation**

$$\{\Delta_A^*, \Delta_B^*, \Delta_R^*\} = \underset{\{\Delta_A, \Delta_B, \Delta_R\}}{\operatorname{argmax}} R_S$$

with  $R_S \triangleq R_A + R_B$ , based on the outer bound

$$R_A \leq \min\{\Delta_A C(\gamma_{AR} + \gamma_{AB}), \Delta_A C(\gamma_{AB}) + \Delta_R C(\gamma_{RB})\}$$

$$R_B \leq \min\{\Delta_B C(\gamma_{BR} + \gamma_{BA}), \Delta_B C(\gamma_{BA}) + \Delta_R C(\gamma_{RA})\}$$

- No sum-rate constraint since **relay is not required to decode**
- Finding  $\Delta_i^*$  amounts to solving a linear program

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# Performance Limits: Symmetric TWRC

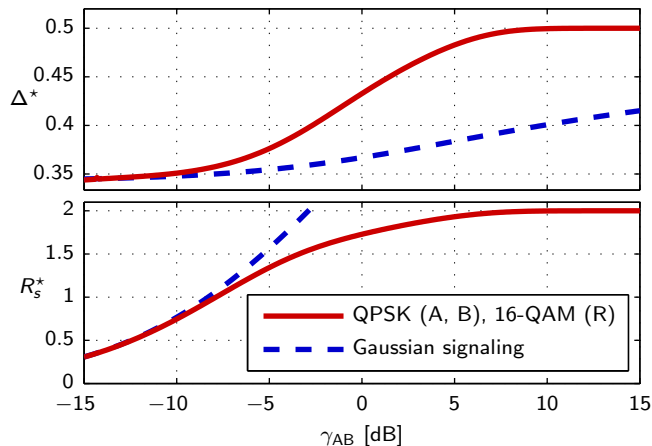
- Restrict to the **symmetric TWRC** with  $d_{AR} = d_{BR} = d_{AB}/2$ 
  - ▶  $\gamma_{AR} = \gamma_{BR}$ ,  $\gamma_{ij} = \gamma_{ji}$ ,
  - ▶  $[\gamma_{AR}]_{\text{dB}} = [\gamma_{BR}]_{\text{dB}} = [\gamma_{AB}]_{\text{dB}} + 10.6 \text{ dB}$
  - ▶  $R \triangleq R_A = R_B$
  - ▶  $\Delta \triangleq \Delta_A = \Delta_B$ ,  $\Delta_R = 1 - 2\Delta$
- Here the **outer bound is tight for DF achievable rates**
- Solving  $\Delta^* = \arg \max_{\Delta} R$  yields

$$\Delta^* = \frac{C_{RB}(\gamma_{RB})}{C_{AR}(\gamma_{AR}) - C_{AB}(\gamma_{AB}) + 2C_{RB}(\gamma_{RB})}$$

if  $C_{AB}(\gamma_{AB}) < \min\{2C_{RB}(\gamma_{RB}), C_{AR}(\gamma_{AR})\}$

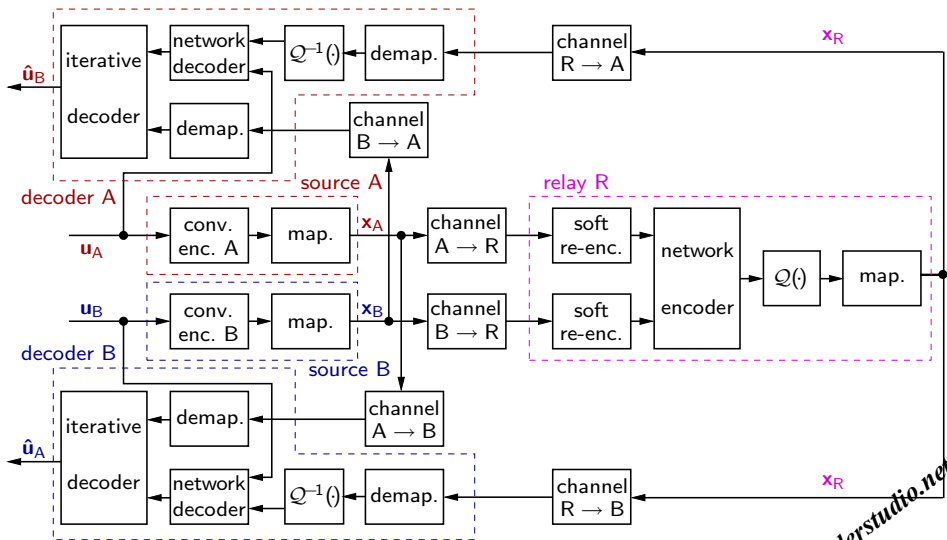
# Performance Limits: Numerical Evaluation

- Example: QPSK map. at A, B; 16 QAM map. at R

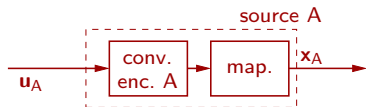


- $\gamma_{AB} \rightarrow \infty \Rightarrow \Delta^* \rightarrow 1/2$  and  $R_s \rightarrow 2$

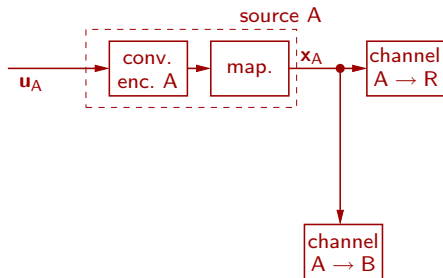
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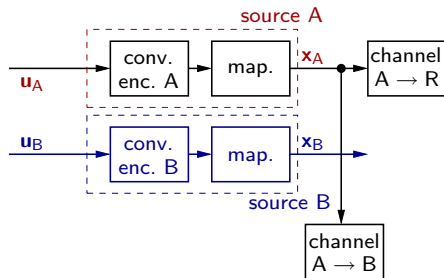
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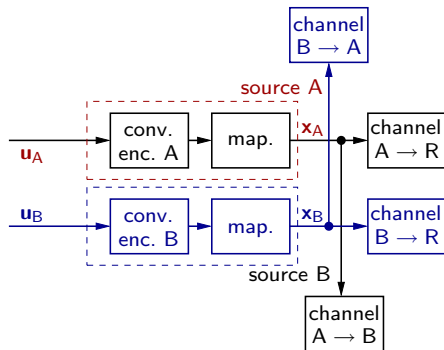
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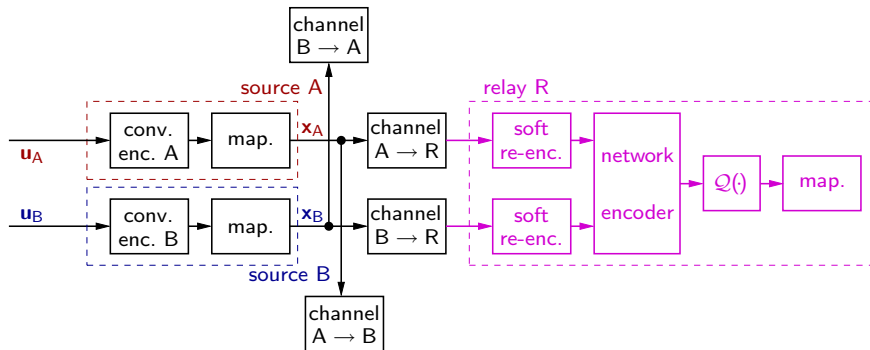
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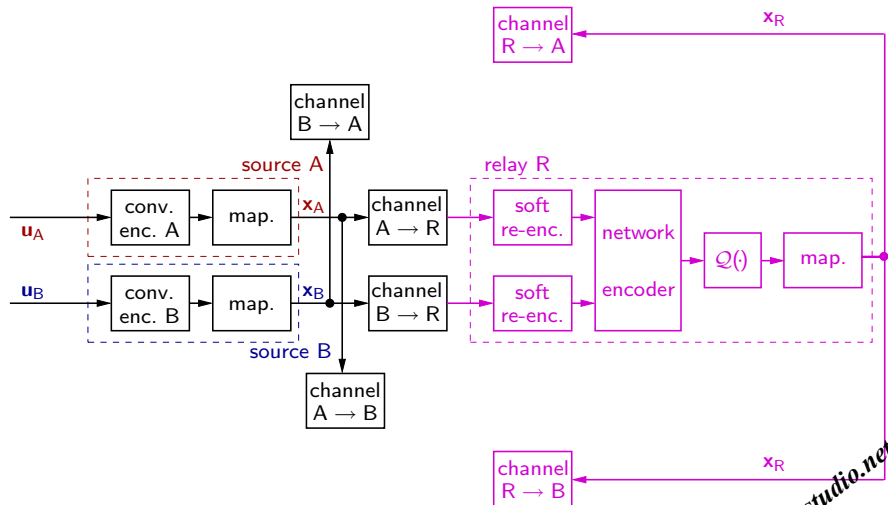
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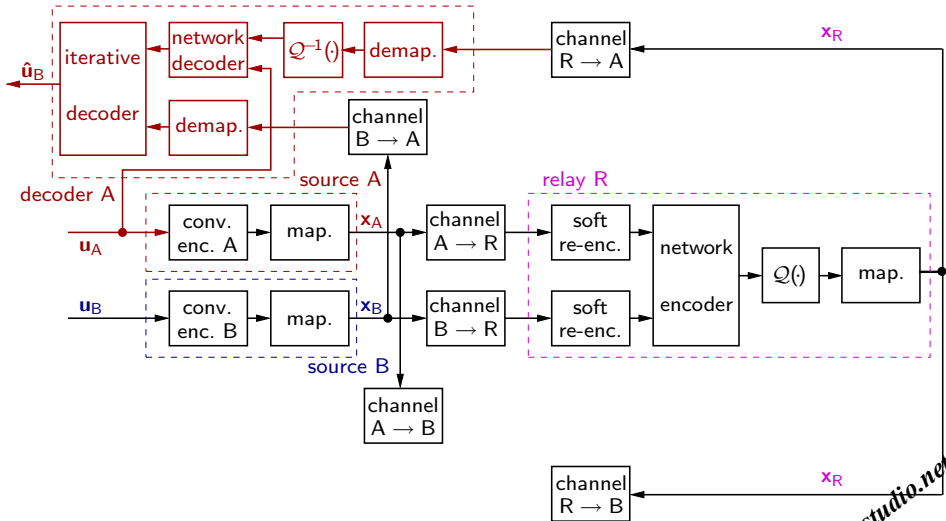
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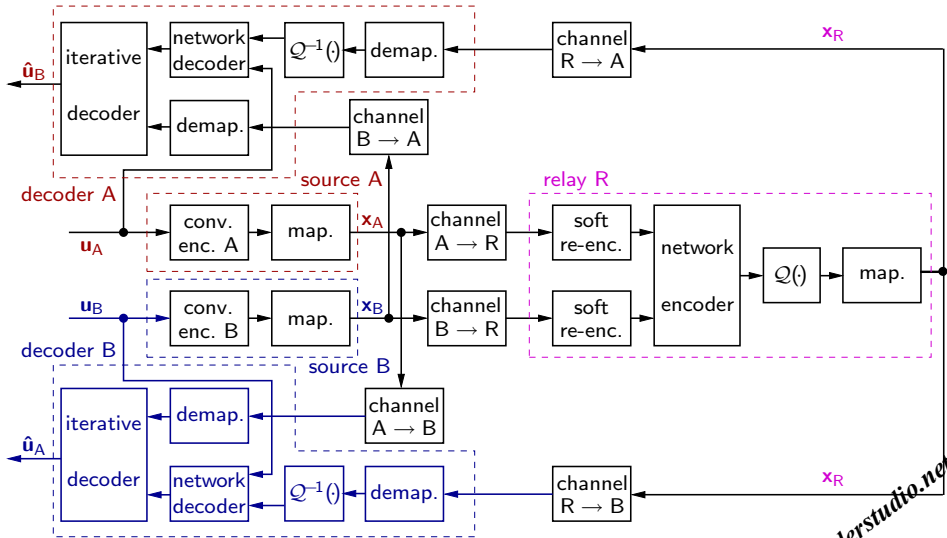
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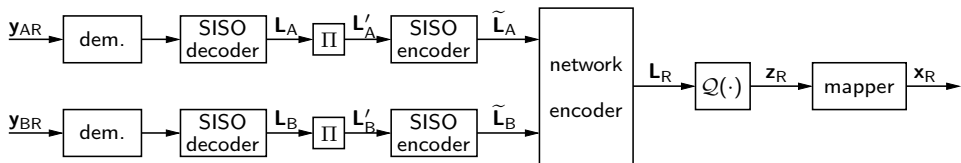
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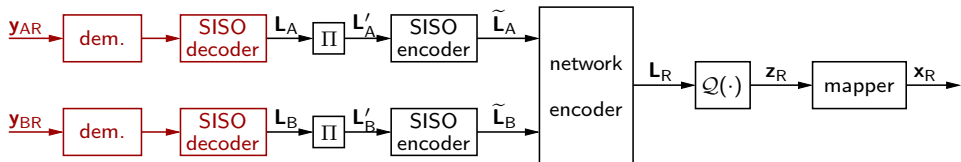


# Relay Operation



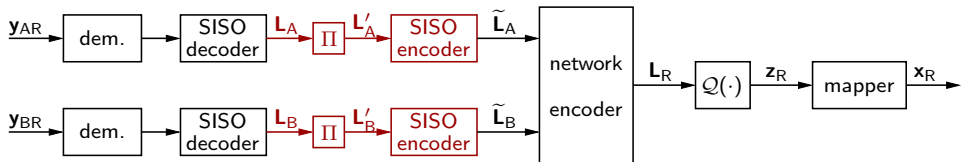
- Demap and decode received signal
- Interleave and re-encode the decoder output
- Compute network-coded L-values  $L_R = \tilde{L}_A \boxplus \tilde{L}_B$
- Quantize  $L_R$  for digital transmission
- Map quantizer output to 16 QAM symbols

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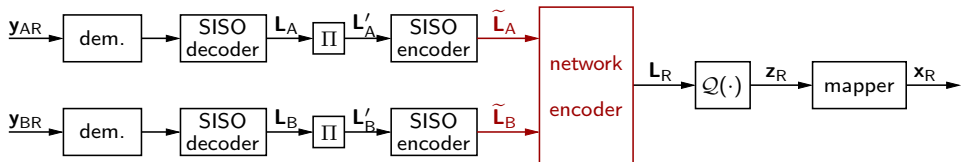
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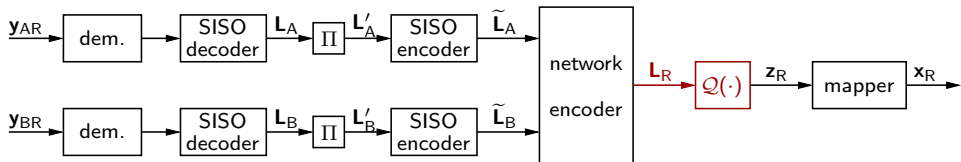
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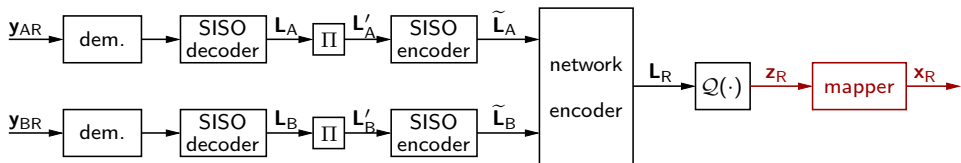
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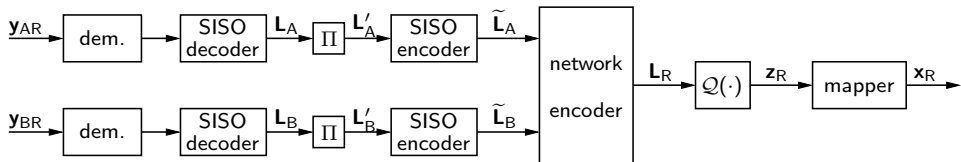
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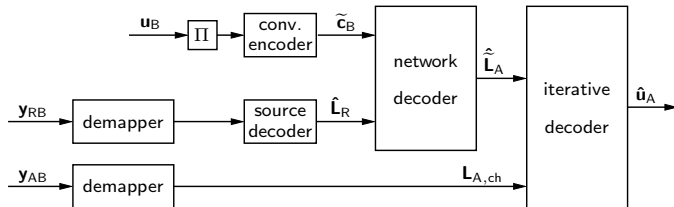


- Remarks

- ▶ **Available information is used even if decoding fails**
- ▶ **Design of quantizer and signal constellation is critical**
- ▶ “Hard” processing possible if relay decoded both signals

# Decoder Operation (1)

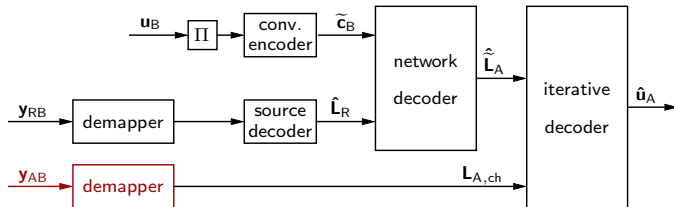
Decoder at node B:



- Demap signal received via direct link
- Reverse network coding performed at relay
  - ▶ source decoder computes an estimate of  $\mathbf{L}_R$
  - ▶ network decoder computes  $\hat{\mathbf{L}}_A = (\mathbf{1} - 2\tilde{\mathbf{c}}_B) \odot \hat{\mathbf{L}}_R$
- Iteratively decode both constituent codes

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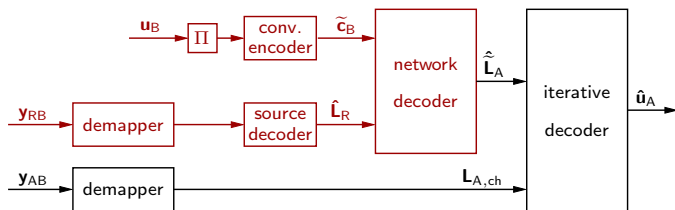
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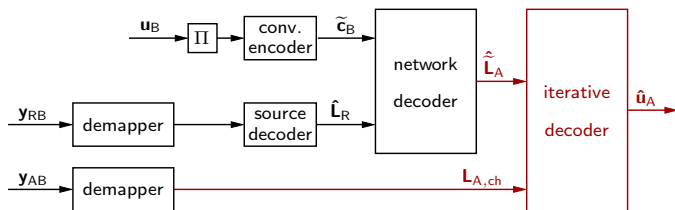
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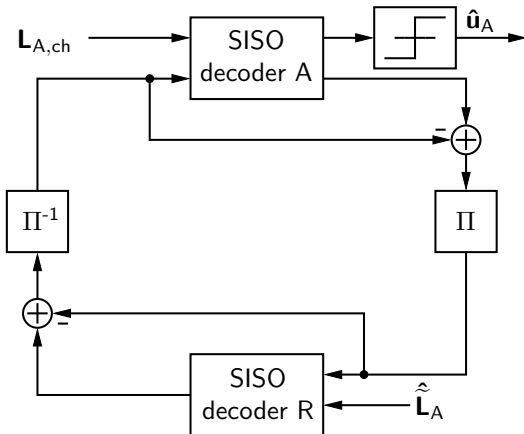
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# Decoder Operation (2)

Iterative “turbo” decoder at node B:



# Quantizer Design (1)

- Design a **scalar Q-level quantizer**  $\mathcal{Q}$  for  $L_R$

$$\mathcal{Q}(L_R) = I_k \quad \text{if} \quad L_R \in \mathcal{I}_k, \quad z_R = k \in \mathcal{Z} = \{0, 1, \dots, Q-1\}$$

- ▶ reproducer value  $I_k$
- ▶ quantization interval  $\mathcal{I}_k$
- Quantization corresponds to **mapping**  $p(z_R|I_R)$
- **Goal:** maximize mutual information  $I(C_R; Z_R)$ ,  $C_R = \tilde{C}_A \oplus \tilde{C}_B$

$$p^*(z_R|I_R) = \arg \max_{p(z_R|I_R)} I(C_R; Z_R)$$

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# Quantizer Design (2)

- One solution: **information bottleneck method** (IBM) [Tishby et al., Allerton99]
  - ▶ iterative algorithm finds **local optimum**
  - ▶ first applied to communications by Zeitler et al.
- Given  $p^*(z_R|I_R)$  **quantizer intervals  $\mathcal{I}_k$  are fixed**
- fix  $l_k$  using L-value of **equivalent discrete channel  $p(z_R|c_R)$**

$$l_k = \log \frac{p(c_R = 0 | z_R = k)}{p(c_R = 1 | z_R = k)}, \quad k \in \mathcal{Z}$$

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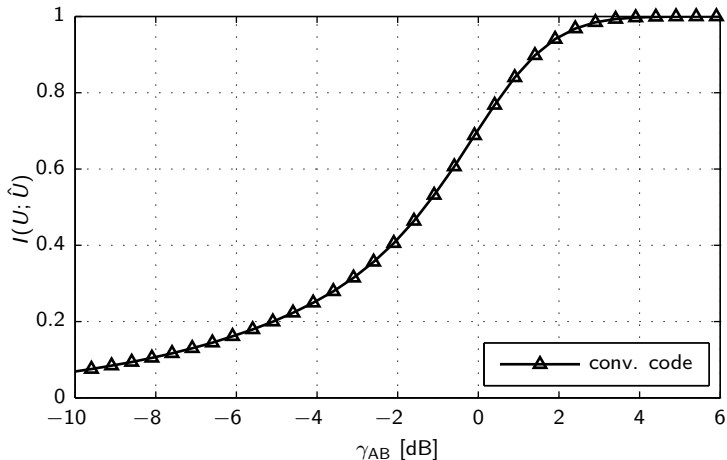
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- Signal constellation and labels have to be designed carefully
- Here **Gray labeling is not optimal**
  - ▶ it matters which bit is in error!
- Use **binary switching** to optimize bit labels and constellation
  - ▶ **eliminates brute-force search**
  - ▶  $(2^4)! \approx 2 \cdot 10^{13}$  possibilities (16 QAM)
  - ▶ **quickly finds a “good” local optimum**
  - ▶ see [ZegerGersho, TrComm90] for details

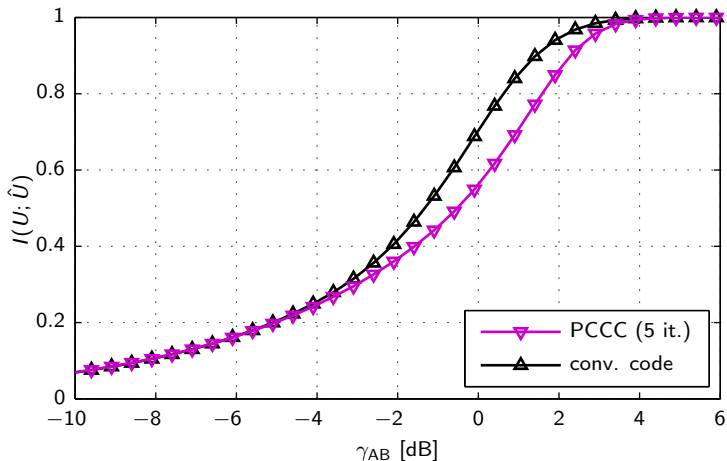
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- $K_A = K_B = 256, R_S = 1, Q \in \{2, 4\}, 1$  decoder iteration



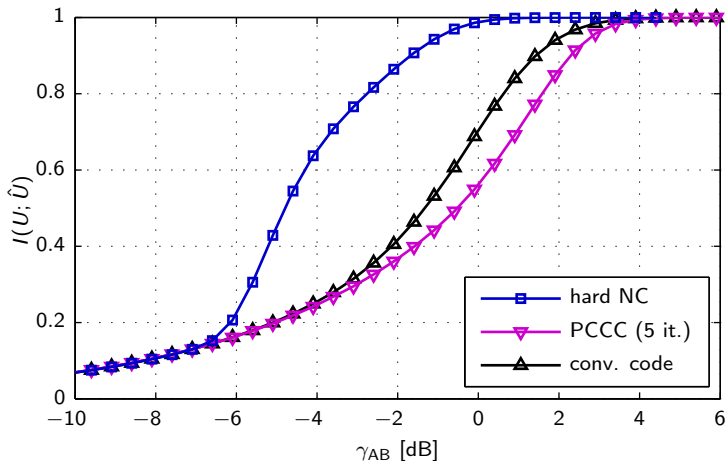
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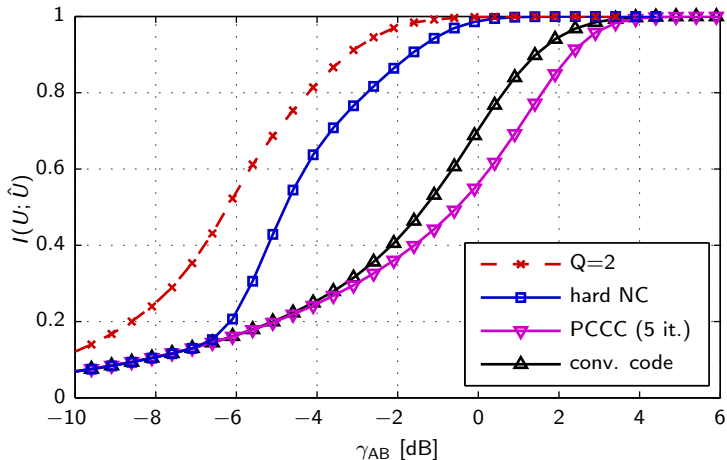
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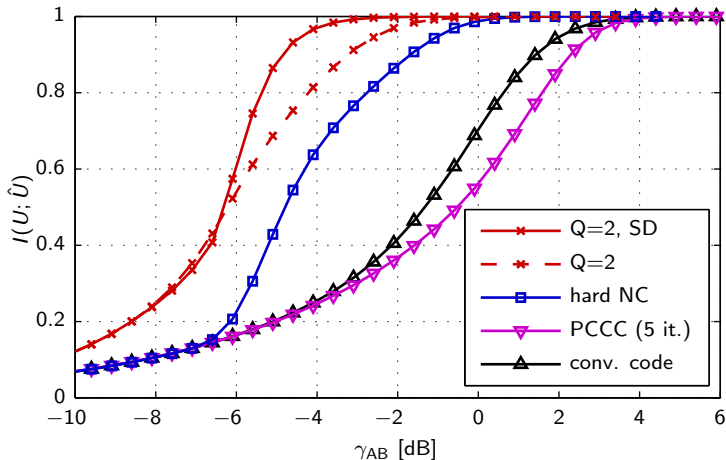
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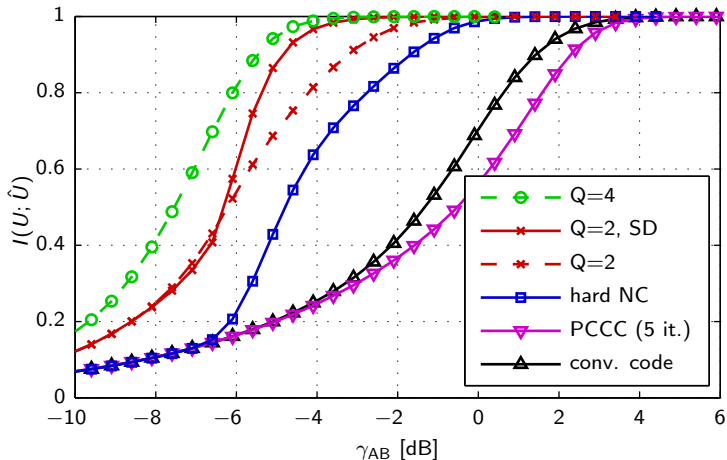
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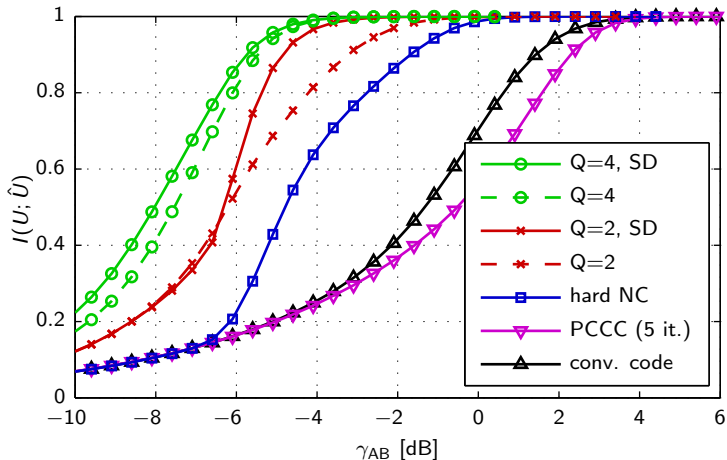
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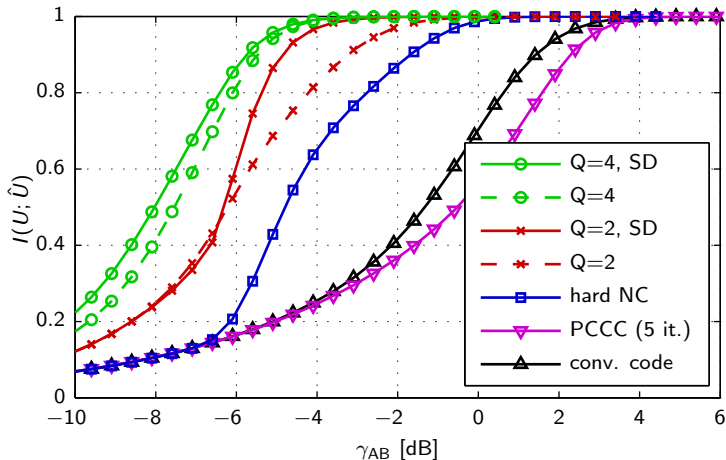
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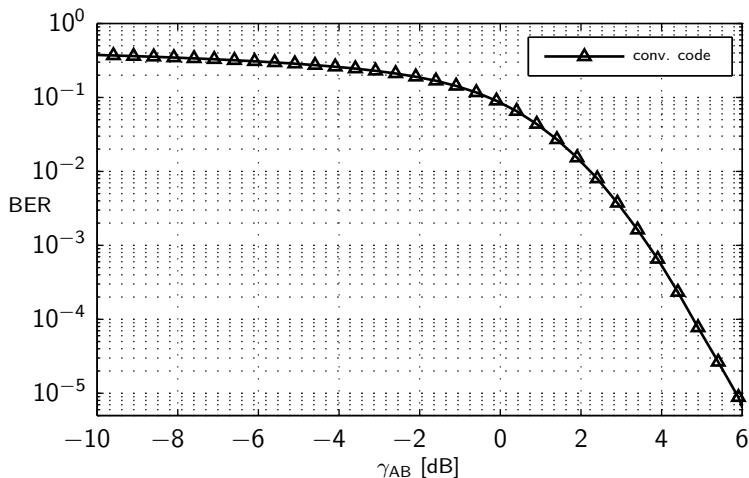
- $K_A = K_B = 256, R_S = 1, Q \in \{2, 4\}, 1$  decoder iteration



- Gain of  $\sim 3.5$  dB over hard NC for a wide range of rates**

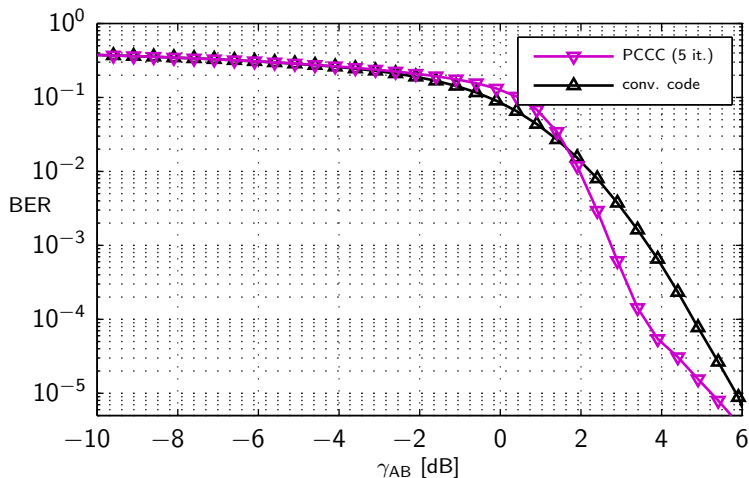
# Simulation Results: Bit Error Ratio

- $K_A = K_B = 256, R_S = 1, Q \in \{2, 4\}, 1$  decoder iteration



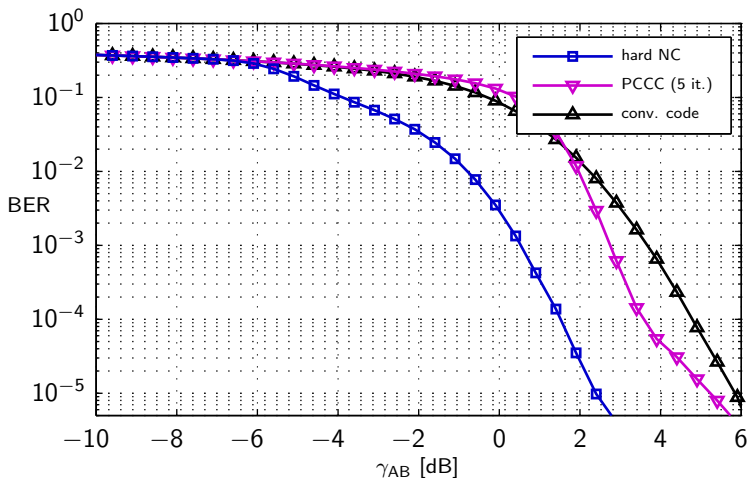
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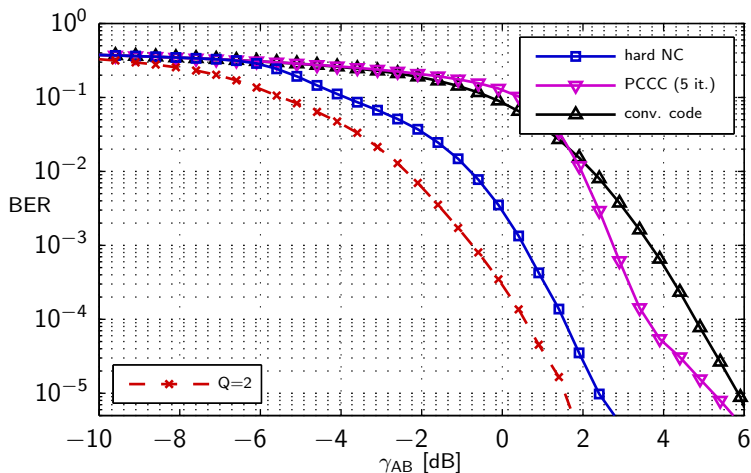
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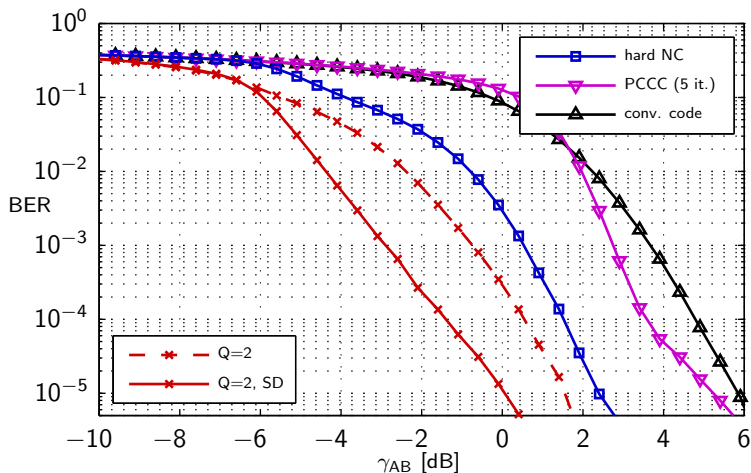
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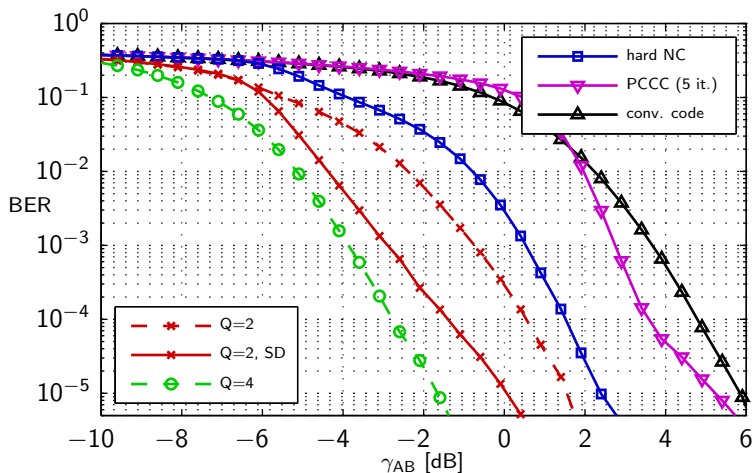
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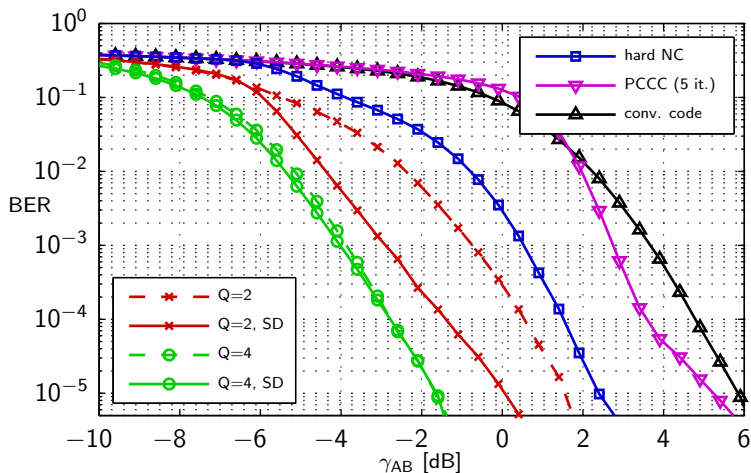
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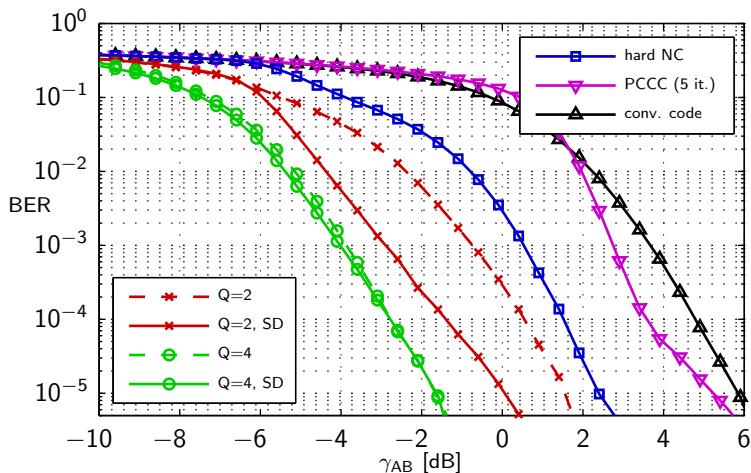
# Simulation Results: Bit Error Ratio

- $K_A = K_B = 256, R_S = 1, Q \in \{2, 4\}, 1$  decoder iteration



# Simulation Results: Bit Error Ratio

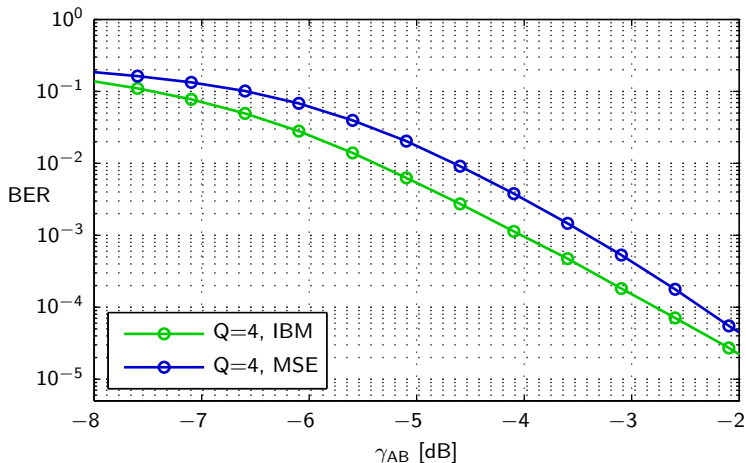
- $K_A = K_B = 256, R_S = 1, Q \in \{2, 4\}, 1$  decoder iteration



- Gain of  $\sim 4.5$  dB over hard NC at  $BER = 10^{-3}$**

# Simulation Results: Quantizer Design

- IBM quantization vs. MSE-optimal quantization ( $Q = 4$ )



- **IBM design gains between 0.5 dB and 1 dB**

## Conclusions:

- Novel soft network coding scheme for the TWRC
- Iterative joint network-channel decoding
- Significant performance gains over “hard” network coding

## Outlook:

- Extension to non-symmetric TWRC
- Data-aided on-the-fly quantizer design
- Code design for block fading channels

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**Thank you!**